On the Optimal Delay Amplification Factor of Multi-Hop Relay Channels

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- Part 1: Motivation & Intuition Motivation A new metric for multi-hop relay channels: The Delay Amplification Factor Preview of main results
- Part 2: Main Results

(Necessary) details of problem set-up Main results & proof sketches Work in progress & Conclusion





Overview

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Low latency communications



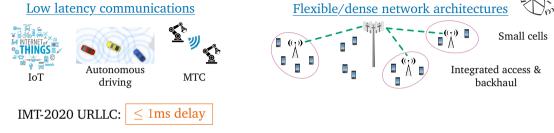


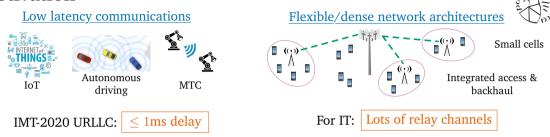
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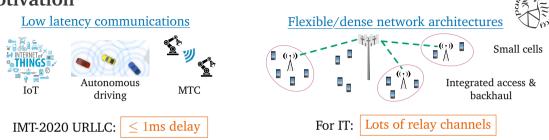




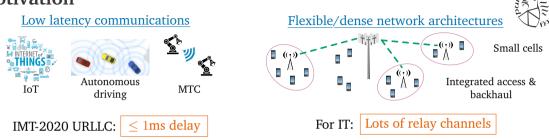




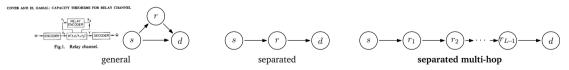


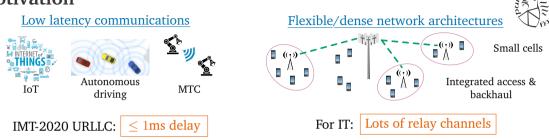


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- ► Interesting from practical perspective: **Separated** relay channel over **multiple hops**

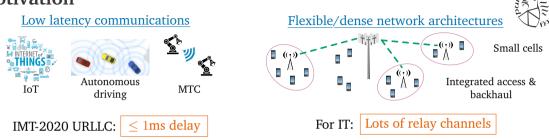




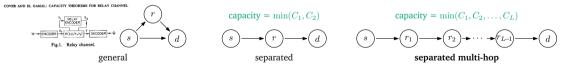
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Capacity analysis is clear, delay-throughput analysis is not

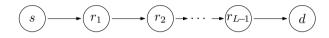


- ► Interesting from practical perspective: **Separated** relay channel over **multiple hops**



- Capacity analysis is clear, delay-throughput analysis is not
- This work: When $R \rightarrow C$, what relaying schemes minimize (relative) delay?







 $\overbrace{C_1}^{\bullet} (r_1) \overbrace{C_2}^{\bullet} (r_2) \xrightarrow{\bullet} \cdots \xrightarrow{\bullet} (r_{L-1}) \overbrace{C_L}^{\bullet} (d)$

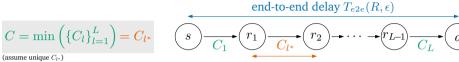


end-to-end delay $T_{e2e}(R,\epsilon)$

$$C = \min\left(\{C_l\}_{l=1}^L\right)$$

$$s) \xrightarrow{} C_1 \xrightarrow{} (r_1) \xrightarrow{} (r_2) \xrightarrow{} \cdots \xrightarrow{} (r_{L-1}) \xrightarrow{} (d)$$

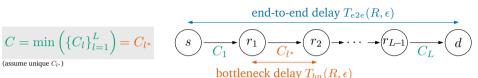




bottleneck delay $T_{bn}(R,\epsilon)$

(assume unique C_{l*})





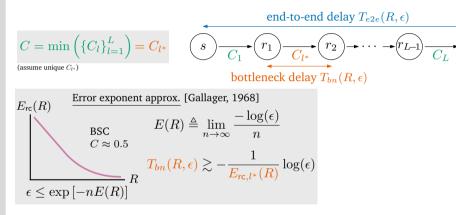
$$\text{DAF}_{\Phi} \triangleq \lim_{R \nearrow C} \lim_{\epsilon \to 0} \frac{T_{e2e}(R, \epsilon)}{T_{bn}(R, \epsilon)}$$

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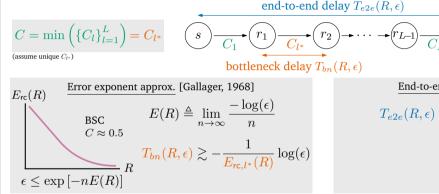




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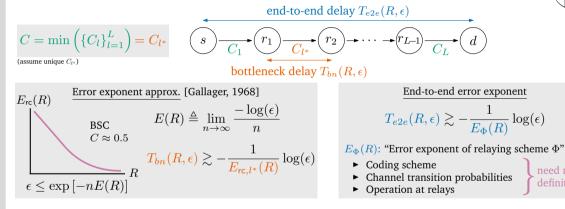
End-to-end error exponent

$$T_{e2e}(R,\epsilon) \gtrsim -\frac{1}{E_{\Phi}(R)}\log(\epsilon)$$

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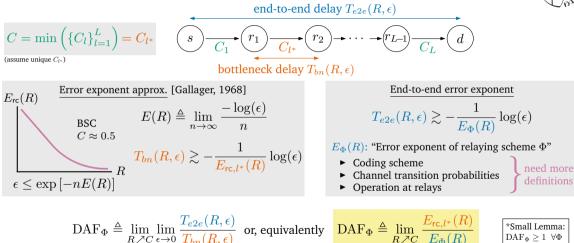
$$E(R) \triangleq \lim_{n \to \infty} \frac{-\log(\epsilon)}{n}$$

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Definition

We define the Delay Amplification factor of an L-hop relaying scheme Φ as $\mathrm{DAF}_{\Phi} \triangleq \lim_{R \neq C} \frac{E_{\mathsf{rc},l^*}(R)}{E_{\Phi}(R)}.$

apple 2 miles

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Lemmas

- 1. Regardless of the relaying scheme, we always have $DAF \ge 1$
- 2. For decode-&-forward schemes, we have $DAF_{DF} = O(L)$

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Main results

- 1. A relaying scheme where $DAF_{\Phi} = 1$ if $l^* = L$
- 2. A 1-bit stop feedback relaying scheme where $DAF_{\Phi'} = 1$ regardless of the position of l^*

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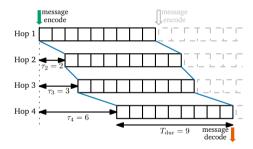
Starting times, relay encoding & joint decoding

A relaying scheme is defined by

- Message $M \in \mathcal{M} \triangleq \{1, \dots, |\mathcal{M}|\}$
- L deterministic starting times $\tau_1 = 0 \le \tau_2 \le \cdots \le \tau_L$
- Transmission duration T_{dur}

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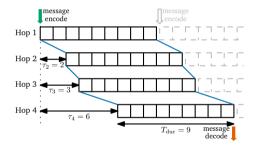




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► Coding scheme

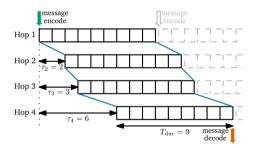
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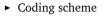
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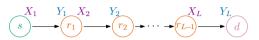
 $Y_1 \, X_2$



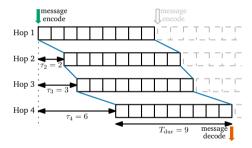
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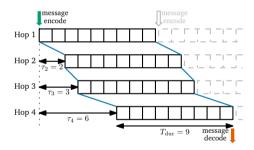


► Block encoding @ source $X_1(t) = f_t^{[1]}(\mathbf{M})$



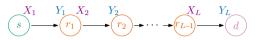
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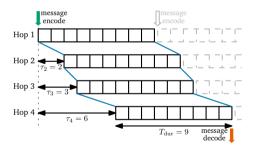




- ► Block encoding @ source $X_1(t) = f_t^{[1]}(\mathbf{M})$
- ► Causal-sequential coding @ relays $Y_l(t) = \text{DMC}_l(X_l(t))$ $X_l(t) = f_l^{[l]}([Y_{l-1}]_{t-1}^{t-1})$

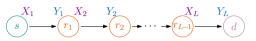
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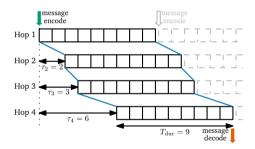




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- Block decoding @ destination
 - $\widehat{M} = g([Y_L]^{\tau_L + T_{\mathrm{dur}}}_*)$

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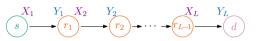
Definition: A (T, R, ϵ) -tuple is <u>achievable</u> by relaying scheme Φ if (i) $T \ge \tau_L + T_{dur}$ (ii) $R \le \frac{\log(|\mathcal{M}|)}{T_{dur}}$ (iii) $\epsilon \ge P(M \ne \widehat{M})$

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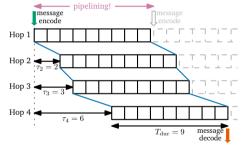
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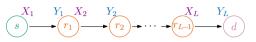
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Block decoding @ destination

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Pipelining:

Source sends next message before previous message is received by destination

Error exponent of relaying scheme Φ

Definition: A (T, R, ϵ) -tuple is achievable by relaying scheme Φ if (i) $T \ge \tau_L + T_{dur}$ (ii) $R \le \frac{\log(|\mathcal{M}|)}{T_{dur}}$ (iii) $\epsilon \ge P(M \ne \widehat{M})$



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(i) T ≥ τ_L + T_{dur}
(ii) R ≤ log(|M|)/T_{dur}
(iii) ε > P(M ≠ M)

• Let $\mathcal{A}_{\phi} \triangleq \{ all (T, R, \epsilon) achievable by \Phi \}$



Error exponent of relaying scheme Φ

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- Let $\mathcal{A}_{\phi} \triangleq \{ all (T, R, \epsilon) achievable by \Phi \}$
- Then define the error exponent of an L-hop relaying scheme Φ as

$$E_{\Phi}(R) \triangleq \limsup_{T \to \infty} \sup_{\epsilon: (T, R, \epsilon) \in \mathcal{A}_{\Phi}} \frac{-\log(\epsilon)}{T}$$



Aside: decode-&-forward, amplify-&-forward



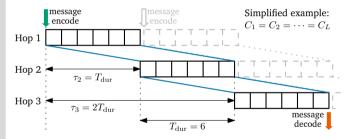
Definitions & analysis framework captures many possible relaying schemes, examples:

Aside: decode-&-forward, amplify-&-forward

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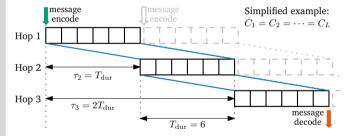
- Random coding over every hop
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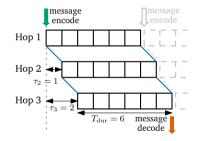
Definitions & analysis framework captures many possible relaying schemes, examples:

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- Random coding between source and destination
- Relays forward received symbols

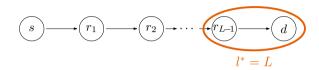


Main Result 1: Feedback-free setting



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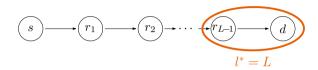


In the feedback-free setting, if the bottleneck hop is the last hop (i.e. $l^* = L$), then transcoding achieves

 $DAF_{TC} = 1 \le DAF_{\Phi}, \ \forall \Phi.$

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Proof includes:

- Description of coding scheme (source, relays, destination)
- ► Derivation of error probability & delay \Rightarrow Error exponent \Rightarrow DAF

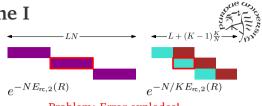
To reduce delay: Break codewords into K micro-blocks, each of length Δ symbols

- Each protected by random coding with rate C_L
- Decode-&-forward over the first L 1 hops



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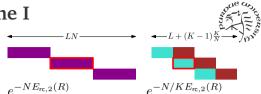
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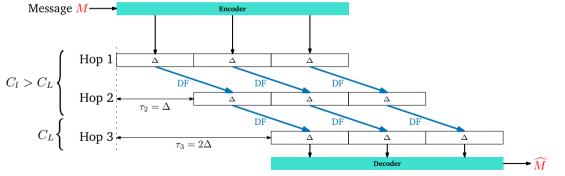
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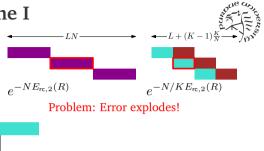
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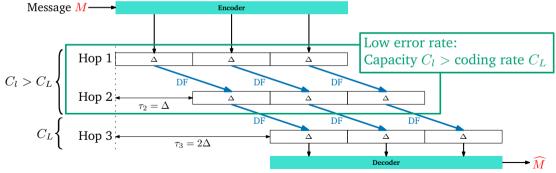


(Example: L = 3, K = 3)

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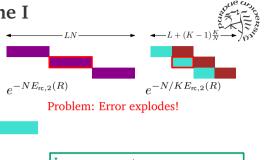


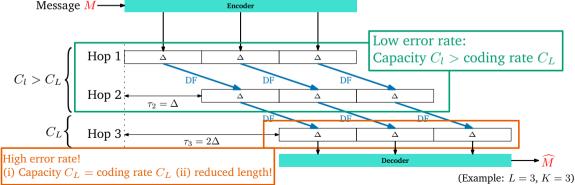


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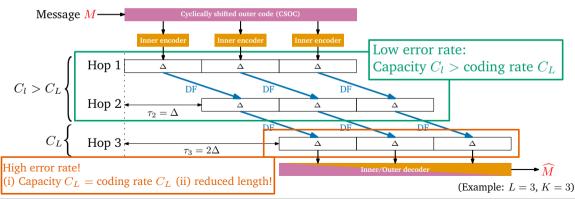
Protect message over last hop using additional coding

- Inspired by concatenated coding
- ► CSOC outer code + random codes for micro-blocks
- Maximum-likelihood decoding at destination

Cyclically shifted outer code (CSOC):

$$\left\{ (\mathbf{c}_{i_1}^{[1]}, \cdots, \mathbf{c}_{i_K}^{[K]}) : \sum_{k=1}^{K} i_k \mod e^{K\Delta(R_I - R)} = 0 \right\}$$

Number of messages: $e^{K\Delta R_I} / e^{K\Delta(R_I - R)} = e^{K\Delta R}$



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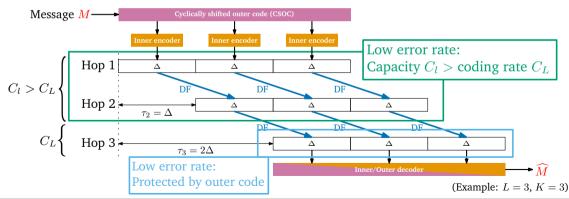
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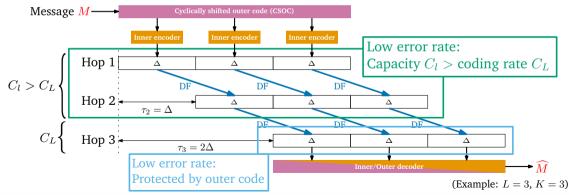
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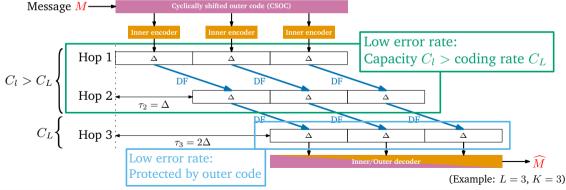
- Choose K such that (i) $K \cdot E_{rec} I(B) < \min$
 - (i) $K \cdot E_{\mathsf{rc},L}(R) < \min_{l \in [1,L-1]} E_{\mathsf{rc},l}(C_L)$ (ii) $K \cdot (C_L - R) \le C_L$ for any $R < C_L$.



- Choose K such that

 (i) K · E_{rc,L}(R) < min_{l∈[1,L-1]} E_{rc,l}(C_L)
 (ii) K · (C_L − R) ≤ C_L for any R < C_L.

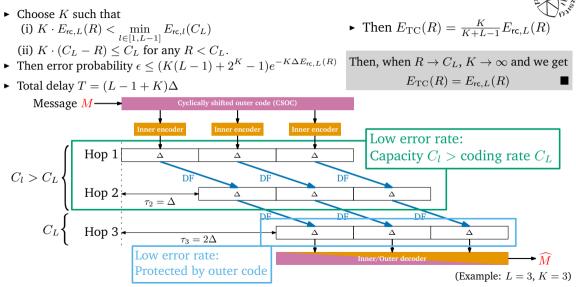
 Then error probability ε ≤ (K(L − 1) + 2^K − 1)e^{-KΔE_{rc,L}(R)}
- Then entry probability $E \leq (R(L-1)+2) = 1$
- Total delay $T = (L 1 + K)\Delta$





 \blacktriangleright Choose K such that • Then $E_{\mathrm{TC}}(R) = \frac{K}{K+L-1}E_{\mathrm{rc},L}(R)$ (i) $K \cdot E_{\mathsf{rc},L}(R) < \min_{l \in [1,L-1]} E_{\mathsf{rc},l}(C_L)$ (ii) $K \cdot (C_L - R) < C_L$ for any $R < C_L$. • Then error probability $\epsilon \leq (K(L-1) + 2^K - 1)e^{-K\Delta E_{rc,L}(R)}$ • Total delay $T = (L - 1 + K)\Delta$ Message $M \longrightarrow$ Cyclically shifted outer code (CSOC) Low error rate: Hop 1 Capacity $C_l > \text{coding rate } C_L$ $C_l > C_L$ Hop 2 DF DF Δ Δ $\tau_2 = \Delta$ DE DE C_L Hop 3 Δ Λ Λ $\tau_3 = 2\Delta$ Low error rate: Protected by outer code (Example: L = 3, K = 3)

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Main Result 2: Stop feedback setting



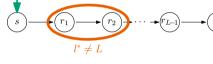
Main Result 2: Stop feedback setting

Some new definitions:

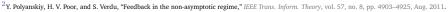
- Allow T_{dur} to be a stopping time, use $E[T_{dur}]$
- Use random coding error exponent with stop feedback [Polyanskiy, Poor, Verdu 2011]

$$E_{\mathsf{sf},l}(R) = (C_l - R)^+$$

• Then
$$\overline{\text{DAF}}_{\Phi} \triangleq \lim_{R \nearrow C_{l^*}} \frac{E_{\mathsf{sf},l}(R)}{E_{\Phi}(R)} = \lim_{R \nearrow C_{l^*}} \frac{C_{l^*} - R}{E_{\Phi}(R)}$$



one-time stop feedback





Main Result 2: Stop feedback setting

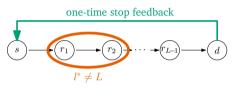
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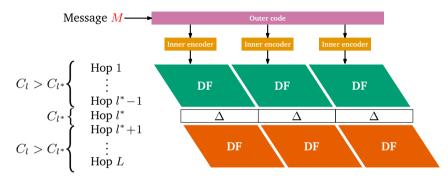


In the stop feedback setting, **regardless of the position of the bottleneck link**, transcoding achieves

$$\overline{\text{DAF}}_{\text{TC}} = 1$$

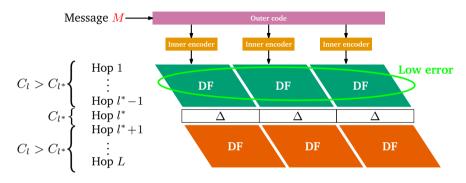
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²Y. Polyanskiy, H. V. Poor, and S. Verdu, "Feedback in the non-asymptotic regime," *IEEE Trans. Inform. Theory*, vol. 57, no. 8, pp. 4903–4925, Aug. 2011.



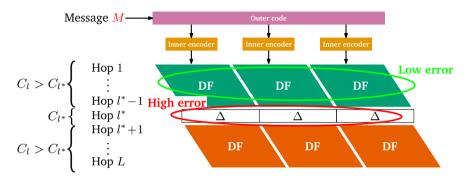
▶ When $l^* \neq L$, all post-bottleneck hops have high error rate due to error propagation





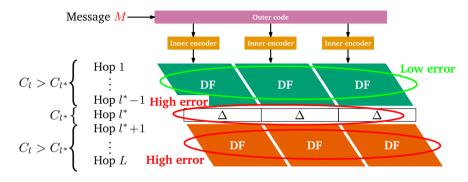
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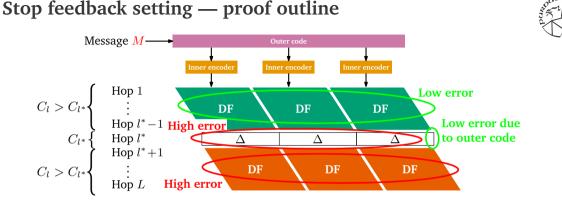


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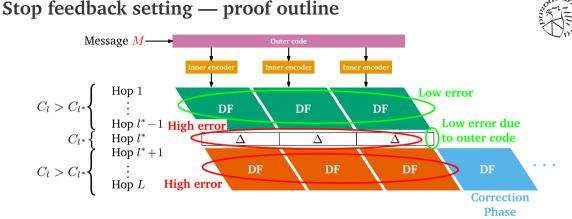




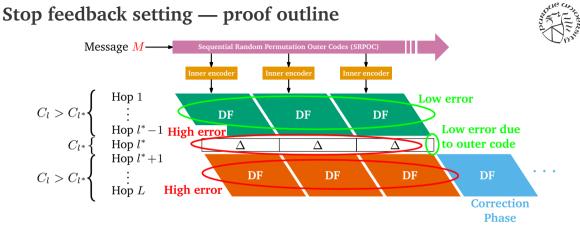
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- ▶ When $l^* \neq L$, all post-bottleneck hops have high error rate due to error propagation
- Bottleneck relay (BR) can decode outer code after K micro-blocks

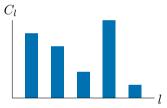


- When $l^* \neq L$, all post-bottleneck hops have high error rate due to error propagation
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- ► Bottleneck relay (BR) can decode outer code after *K* micro-blocks
- ► BR sends additional redundancy until destination signals stop feedback
- \blacktriangleright With careful code construction can show $\overline{\rm DAF}_{\rm TC}=1$

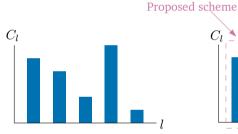




With stop feedback: Solved Feedback-free: Solved

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With stop feedback: Solved Feedback-free: Solved

With stop feedback: Solved Feedback-free: Partially solved

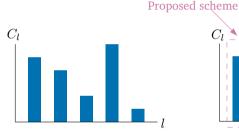
DF over

groups

 C_{l}

Proposed scheme





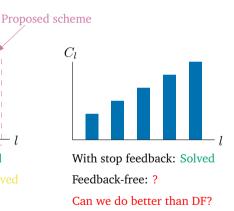
With stop feedback: Solved Feedback-free: Solved

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groups

 C_1

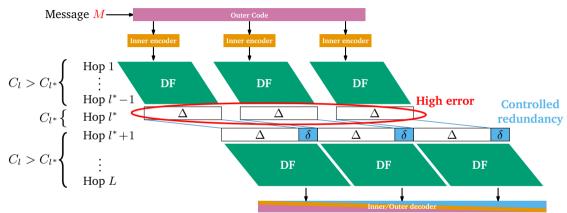


What is left? — Case $l^* \neq L$ without feedback



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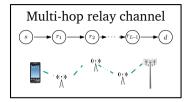


- ► BR adds redundancy per micro-block
- Conjecture: $1 < DAF_{TC'} < DAF_{DF}$

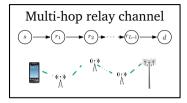
Conclusion





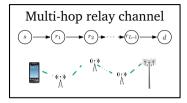






When $R \rightarrow C$, what relaying schemes minimize relative delay?



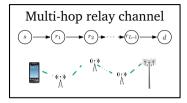


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$$s \xrightarrow{C_1} (r_1) \xrightarrow{C_{l^*}} (r_2) \xrightarrow{\bullet} \cdots \xrightarrow{\bullet} (r_{L-1}) \xrightarrow{C_L} (d)$$

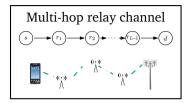
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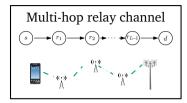


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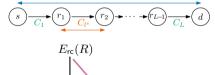
$$s \xrightarrow{} c_1 \xrightarrow{} c_2 \xrightarrow{} \cdots \xrightarrow{} c_L \xrightarrow{} d$$

Delay Amplification Factor $DAF_{\Phi} \triangleq \lim_{R \nearrow C} \frac{E_{\mathsf{rc},l^*}(R)}{E_{\Phi}(R)}$





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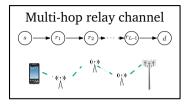




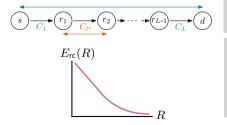
Delay Amplification Factor $DAF_{\Phi} \triangleq \lim_{R \neq C} \frac{E_{\mathsf{rc},l^*}(R)}{E_{\Phi}(R)}$

Main Takeaways

- ► Relaying schemes with DAF = 1 are possible
- Linearly growing delay of decode-&-forward is NOT a fundamental limit



When $R \rightarrow C$, what relaying schemes minimize relative delay?





Delay Amplification Factor $DAF_{\Phi} \triangleq \lim_{R \neq C} \frac{E_{\mathsf{rc},l^*}(R)}{E_{\Phi}(R)}$

Main Takeaways

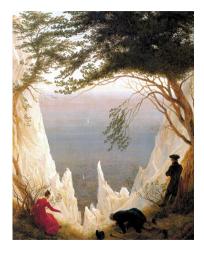
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Future Work

- $l^* \neq L$ without feedback
- Applications to Gaussian, Rayleigh fading channels



Questions.



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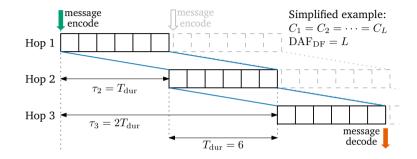
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Backup Slides.

Aside: DAF of decode-&-forward



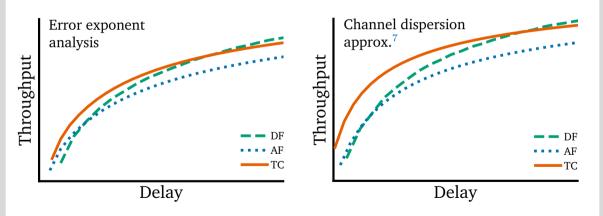


- Operate each hop close to capacity $\Rightarrow T \approx \sum_{l} \frac{B}{C_{l}}$ for a *B*-bit message
- But $T_{bn} \approx \frac{B}{C_{l^*}}$
- Then $\text{DAF}_{\text{DF}} = \frac{T}{T_{bn}} = \sum_{l} \frac{C_{l^*}}{C_l} > 1$

N.B.: AF does not achieve capacity, DAF does not apply

Prior Work: Transcoding





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